

Task 7 (SISO systems assumed) 1/2

a) An input-output polynomial model

$$A(q)y_k = B(q)u_k + C(q)e_k$$

where $A(q)$, $B(q)$ and $C(q)$ are polynomials in the shift operator q^{-1} . (ARIMAX)

Auto Regressive Moving Average with extra (exogenous) variables

b) An input-output polynomial model

$$A(q)y_u = B(q)u_k + e_u$$

Auto Regressive with extra variables - ARX

c) A deterministic model

$$\begin{aligned} x_{n+1} &= Ax_n + Bu_n \\ y_n &= Dx_n \end{aligned}$$

may be written as an ARX model

d) A more general state space model

$$\left\{ \begin{array}{l} x_{n+1} = Ax_n + Bu_n \\ y_n = Dx_n + e_n \end{array} \right.$$

may be written as an ARIMAX model.

e)

e)

$$x_n = a x_{n-1} + b u_{n-1} + k e_{n-1}$$

and $x_n = g_n - e_n$ gives

$$\underline{g_n - e_n} = a(\underline{g_{n-1} - e_{n-1}}) + \underline{b u_{n-1}} + k e_{n-1}$$

$$g_n - a g_{n-1} = b u_{n-1} + e_n + (k-a) e_{n-1}$$

gives $\frac{A(q)}{(1-aq^{-1})} g_n = \underbrace{b q^{-1} u_n}_{B(q^{-1})} + \underbrace{(1-(k-a)q^{-1}) e_n}_{C(q^{-1})}$

where $q^{-1} g_n = g_{n-1}$, $q^{-1} u_n = u_{n-1}$, and $q^{-1} e_n = e_{n-1}$.

With $k=a$ we have $C(q^{-1})=1$ and an

ARX model

f)

⇒

7f)

$$x_n = \theta_1 x_{n-1} + \theta_2 u_{n-1} + \theta_3 e_{n-1}$$

and $x_n = y_n - e_n$ gives

$$\underline{y_n - e_n} = \theta_1 (\underline{y_{n-1} - e_{n-1}}) + \theta_2 \underline{u_{n-1}} + \theta_3 \underline{e_{n-1}}$$

$$y_n = \theta_1 y_{n-1} + \theta_2 u_{n-1} + e_n = [y_{n-1} \ u_{n-1}] \begin{bmatrix} \theta_1 \\ \theta_2 \end{bmatrix} + e_n$$

We write as linear regression model

$$y_n = \varphi_n^T \theta + e_n$$

$$\text{where } \varphi_n^T = [y_{n-1} \ u_{n-1}], \theta = \begin{bmatrix} \theta_1 \\ \theta_2 \end{bmatrix}$$

7g)

$$\hat{\theta}_N = \left(\sum_{k=1}^N \varphi_k \varphi_k^T \right)^{-1} \sum_{k=1}^N \varphi_k y_k$$

$$\text{assuming } \varepsilon_k = y_k - \hat{y}_k(\theta) = y_k - \varphi_k^T \theta$$

• When $y_n = \varphi_n^T \theta + e_n$ and $E(e_n e_n^T) = \Delta$.
Optimal weight matrix

$$\Lambda = \Delta^{-1} = (E(e_n e_n^T))^{-1}$$

i.e. The BLUE, Best Linear Unbiased Estimate

Task 2

a) Given $L=3$, $Y=4$ and

$$H_k = DA^{k-1}B \quad \forall k=1, \dots, 7. \quad N=L+Y=7$$

- $H_{2|4} = \begin{bmatrix} H_2 & H_3 & H_4 & H_5 & H_6 \\ H_3 & H_4 & H_5 & H_6 & H_7 \end{bmatrix}, H_{1|4} = \begin{bmatrix} H_1 & H_2 & H_3 & H_4 & H_5 \\ H_2 & H_3 & H_4 & H_5 & H_6 \end{bmatrix}$
- $L=2$ number of ~~states~~ ^{states} in $H_{1|2}$ and $H_{2|2}$
 $Y=4 \rightarrow 11$ - of columns in $\rightarrow 11$
- $H_{1|L} = O_L C_J$ and $H_{2|L} = O_L A C_J$

where

$$O_L = \begin{bmatrix} D \\ DA \\ \vdots \\ DA^{L-1} \end{bmatrix}$$

- the extended observability matrix.

$$C_J = [B \quad AB \quad \cdots \quad A^{Y-1}B] \quad \begin{cases} \text{The extended} \\ \text{controllability} \\ \text{matrix} \end{cases}$$

- Related to \det of L

System order bounded by

$$1 \leq n \leq L \cdot m$$

where $\dim(G_n) = m$, $D \in \mathbb{R}^{m \times n}$

2b) Initial value for $\bar{X}_{k=0}$ at startup. Constant K .

$$\bar{y}_k = D\bar{X}_k + EU_k \quad - \text{Predicted output}$$

$$E_k = y_k - \bar{y}_k \quad - \text{Innovations}$$

$$\hat{\bar{X}}_k = \bar{X}_k + K E_k \quad - \begin{array}{l} \text{Aposteriori estimate} \\ (\text{K-Kalman gain}) \end{array}$$

$$\bar{X}_{k+1} = A\hat{\bar{X}}_k + BU_k \quad - \text{update a priori state estimate}$$

Varying Kalman gain matrix or constant

$$K = \bar{X}D^T(D\bar{X}D^T + W)^{-1}$$

where $\bar{X} = E((x - \bar{x})(x - \bar{x})^T)$

May eliminate $\hat{\bar{X}}$ and we obtain

$$\left. \begin{aligned} \bar{X}_{k+1} &= A\bar{X}_k + BU_k + \hat{K}E_k \\ y_k &= D\bar{X}_k + EU_k + E_k \end{aligned} \right\} (1)$$

where $\hat{K} = AK$

the Kalman gain matrix is innovations form (1).

- SVD of $H_{11L} = O_L G$

$$H_{11L} = U S V^T = [U, V] \begin{bmatrix} S_1 & 0 \\ 0 & \ddots \\ 0 & 0 & S_2 \end{bmatrix} \begin{bmatrix} V, V_2 \end{bmatrix}^T$$

$$= O_L S_1 V_1^T$$

- system order $n =$ the number of non-zero singular values

$$S_1 = \begin{bmatrix} S_1 & 0 \\ 0 & \ddots \\ 0 & \cdots & S_n \end{bmatrix}, S_2 \approx 0$$

- $O_L = U_1 S_1$ and $G_J = V_1^T$
gives output normal realization

- Solve A from $H_{21L} = O_L A G_J$
gives

$$A = (O_L^T O_L)^{-1} O_L^T H_{21L} G_J^T (G_J G_J^T)^{-1}$$

or using SVD, solve $H_{21L} = U, S, A V^T$
 $U^T H_{21L} V_1 = S, A \Rightarrow A = S^{-1} U^T H_{21L} V_1$

2c) For simplicity, constant K

Initial predicted state $\bar{x}_{k=0}$

$$(1) \quad \begin{cases} \bar{g}_n = g(\bar{x}_n) \\ \hat{x}_n = \bar{x}_n + K \epsilon_n, \epsilon_n = g_n - \bar{g}_n \\ \bar{x}_{n+1} = f(\hat{x}_n, u_n) \end{cases}$$

- with varying K , change (1) above
- initial values for $\bar{x}_{k=0}$ and $\hat{x}_{k=0}$

$$\bar{g}_n = g(\bar{x}_n)$$

$$D_n = \frac{d g(x_n, u_n)}{dx_n^T} \Big|_{\bar{x}_n, u_n}$$

$$K_n = \bar{x}_n D_n^T (D_n \bar{x}_n D_n^T + W)^{-1}$$

$$\hat{x}_n = \bar{x}_n + K_n \epsilon_n, \epsilon_n = g_n - \bar{g}_n$$

$$\bar{x}_{n+1} = f(\hat{x}_n, u_n)$$

$$\hat{x}_n = (I - K_n D_n) \bar{x}_n (I - K_n D_n)^T + K_n W K_n^T$$

$$\bar{x}_{n+1} = A \hat{x}_n A^T + V$$

Here $V = E(\phi_n \phi_n^T)$ and $W = E(a_n w_n^T)$
 the covariance matrices of ϕ_n and a_n , respectively

2a) Kalman Filter on Prediction form

- $\bar{y}_k = D\bar{x}_k + Eu_k$

$$e_k = y_k - \bar{y}_k$$

$$\bar{x}_{k+1} = Ax_k + Bu_k + \tilde{k} \underbrace{(y_k - \bar{y}_k)}_{e_k}$$

or

$$\begin{aligned}\bar{x}_{k+1} &= \bar{x}_k + Bu_k + k(y_k - \bar{y}_k) \\ \bar{y}_k &= Dx_k + Eu_k\end{aligned}$$

Prediction Error (PE)

$$e_k = y_k - \bar{y}_k = y_k - D\bar{x}_k - Eu_k$$

$n=2$

$$A = \begin{bmatrix} 0 & 1 \\ \theta_1 & \theta_2 \end{bmatrix}, B = \begin{bmatrix} \theta_3 \\ \theta_4 \end{bmatrix}, K = \begin{bmatrix} \theta_5 \\ \theta_6 \end{bmatrix}$$

$$D = [1 \ 0], E = \theta_7, \bar{x}_{k=0} = \begin{bmatrix} \theta_8 \\ \theta_9 \end{bmatrix}$$

$$\theta = [\theta_1 \ \theta_2 \dots \theta_9]^T$$

Assuming SISO system!

Task 3

a) $\bullet Y_{0/L} = O_L X_0$

where

$$Y_{0/L} = \begin{bmatrix} g_0 & \ddots & x \\ g_1 & \ddots & x \\ \vdots & & \\ g_{L-1} & g_L & \cdots & g_{N-2} \end{bmatrix}$$

$$X_0 = [x_0 \ x_1 \ \cdots \ x_{N-2}]$$

• $\bullet Y_{1/L} = O_L A X_0$

where

$$Y_{1/L} = \begin{bmatrix} g_1 & \ddots & x \\ g_2 & \ddots & x \\ \vdots & & \\ g_L & g_{L+1} & \cdots & g_{N-1} \end{bmatrix}$$

• - A - the transition (system) matrix.

- Eigenvalues of A defines stability, time constants.

b) O_L and X_0 estimated from SVD of $Y_{0/L} = U, S, V^T$ and $O_L = U, S$, and $X_0 = V$. System order $n = \#$ of non zero singular values of $Y_{0/L}$.

c) Solve $Y_{1/L} = U, S, A V^T \Rightarrow A = S^{-1} U^T Y_{1/L} V$

c) or solve

$$Y_{IL} = O_L A X_0$$

$$\text{or } A = O_L^+ Y_{IL} X_0^+$$

where $O_L^+ = (O_L^T O_L)^{-1} O_L^T$ - left pseudo inverse

$$X_0^+ = X_0^T (X_0 X_0^T)^{-1}$$
 - right pseudo inverse

- d) initial state vector X_0 , 1st column, in $X_0 = V^T = [x_0 \ x_1 \ \dots]$
- first block row in $O_L = U S$,

Task 4

a) $N=10, L=2, J=2$, $g=1, \begin{cases} u_k \\ y_k \end{cases} \forall k=0, 1, \dots, 9$ 10/12

Remark, $J=2$ not used in Eq. (22)-(23)
and not used in information

$$Y_{1|L} = \begin{bmatrix} y_1 & y_2 & \dots & y_8 \\ y_2 & y_3 & \dots & y_9 \end{bmatrix}, Y_{0|2} = \begin{bmatrix} y_0 & y_1 & \dots & y_2 \\ y_1 & y_2 & \dots & y_8 \end{bmatrix}$$

$$U_{0|L+g} = U_{0|3} = \begin{bmatrix} u_0 & \dots & u_7 \\ u_1 & \dots & u_8 \\ u_2 & u_3 & \dots & u_9 \end{bmatrix}$$

Remark

$k=8$ columns!

$$O_L = \begin{bmatrix} P \\ DA \end{bmatrix}, H_L^d = \begin{bmatrix} E & 0 \\ DB & E \end{bmatrix}$$

$$\tilde{A}_L = O_L A (O_L^T O_L)^{-1} O_L^T$$

$$\tilde{B}_L = [O_L B \quad H_L^d] - \tilde{A}_L [H_L^d \quad 0]$$

2b) Projections

$$\underbrace{Y_{0/L} V_{0/L+g}}_{Z_{0/L}} = O_L \tilde{x}_0$$

where $Z_{0/L} = Y_{0/L} V_{0/L+g}^\perp$

and $V_x^\perp = I_{KxK} - V_x^T (V_x V_x^T)^+ V_x$ $\left\{ \begin{array}{l} x=0/L+g \\ \text{for} \\ \text{simplification} \\ \text{of notation} \end{array} \right.$
 is such that $V \cdot V^\perp = 0$

$$VV^\perp = V(I - V^T (V V^T)^+ V) = V - V = 0 \quad \underline{\text{qed}}$$

$$Z_{1/L} = Y_{1/L} V_{0/L+g}^\perp$$

c) Solve A from

$$Z_{1/L} = \underbrace{O_L A O_L^+}_{A_L} \cdot Z_{0/L}$$

B and E solved from \tilde{B}_L .

$$\begin{aligned} \tilde{B}_L &= (Y_{1/L} - A_L Y_{0/L}) V_{0/L+g}^\perp \\ &= [X \quad \begin{matrix} 0 \\ \text{E} \end{matrix}] \rightarrow E \end{aligned}$$

- d) In a first step we identify the innovations noise process
 $[e_j e_{j+1} \dots e_{N-1}] = \mathcal{E}$
 by projecting \mathcal{E} past data $\begin{bmatrix} u_{01j} \\ y_{01j} \end{bmatrix}$ onto future outputs in Y_{j+1}

We have

$$\underbrace{Y_{j+1}}_{\tilde{Z}_{j+1}} / \begin{bmatrix} u_{01j} \\ y_{01j} \end{bmatrix} = DX_{j+1} \left\{ \begin{array}{l} \text{Signal} \\ \text{part} \end{array} \right.$$

when $A/B \stackrel{\text{det}}{=} AB^T(BB^T)^+$

$$\tilde{Z}_{j+1}^s = Y_{j+1} - Y_{j+1} / \begin{bmatrix} u_{01j} \\ y_{01j} \end{bmatrix} \left\{ \begin{array}{l} \text{Noise,} \\ \text{innovations} \\ \text{part.} \end{array} \right.$$

- When the noise part is known we solve a deterministic subspace identification problem

$$\left. \begin{array}{l} x_{n+1} = Ax_n + \hat{B}\hat{u}_n \\ \hat{y}_n = Dx_n \end{array} \right\} \begin{array}{l} \hat{u}_n \\ \hat{y}_n \end{array} \right\} \begin{array}{l} \text{Known} \\ \forall k=j, \dots, N-1 \end{array}$$